



Chapter 6: Waveguide

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6.1 Introduction :

❖ **Transmission line:**

Two conductors

$$f < f_{\text{cut}}$$

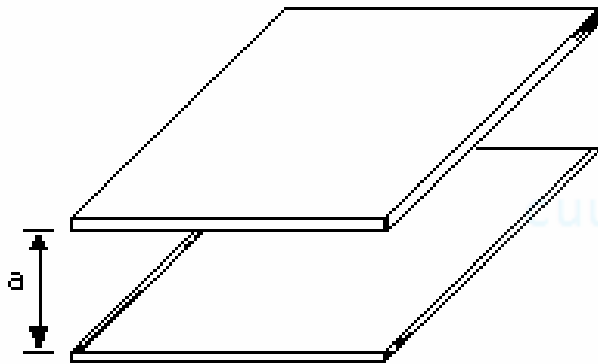
TEM wave

❖ **Waveguide:**

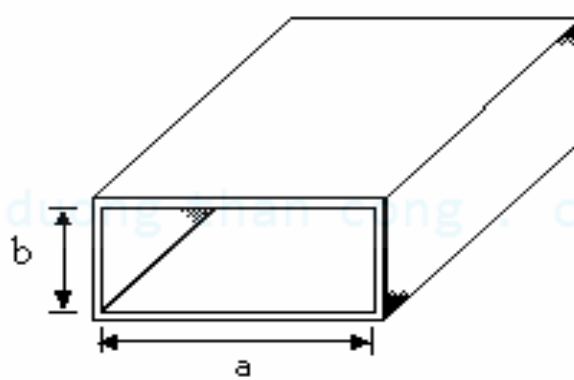
Only one hollow conductors

$$f > f_{\text{cut}}$$

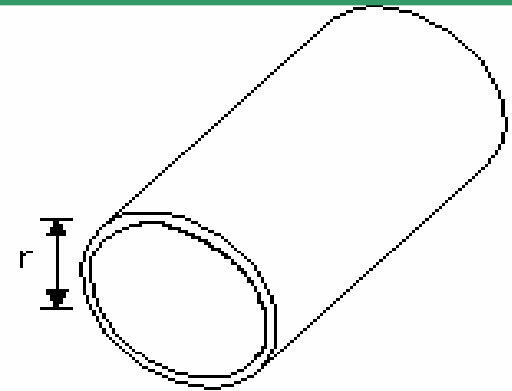
TE and TM waves



a) Parallel-plate waveguide

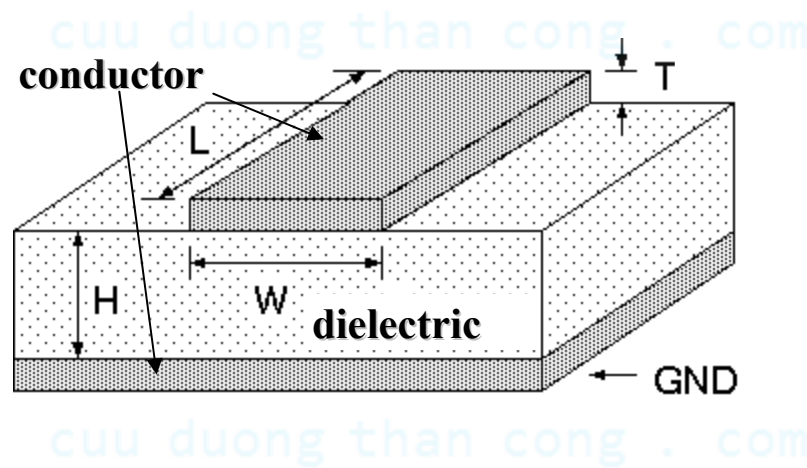
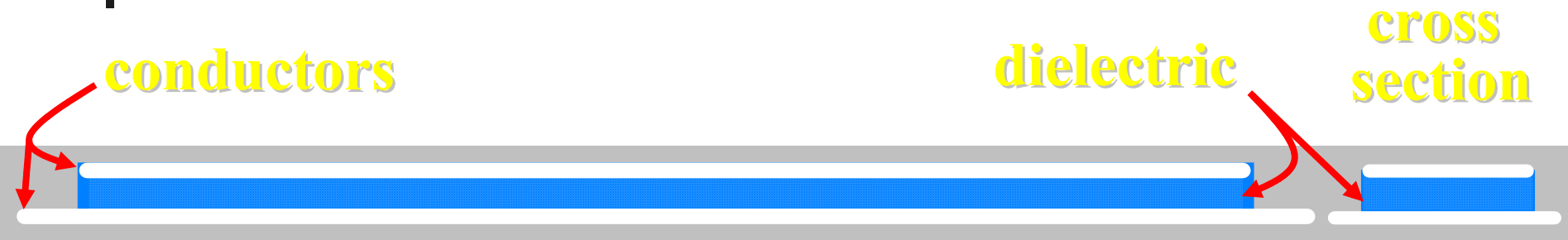


b) Rectangular waveguide

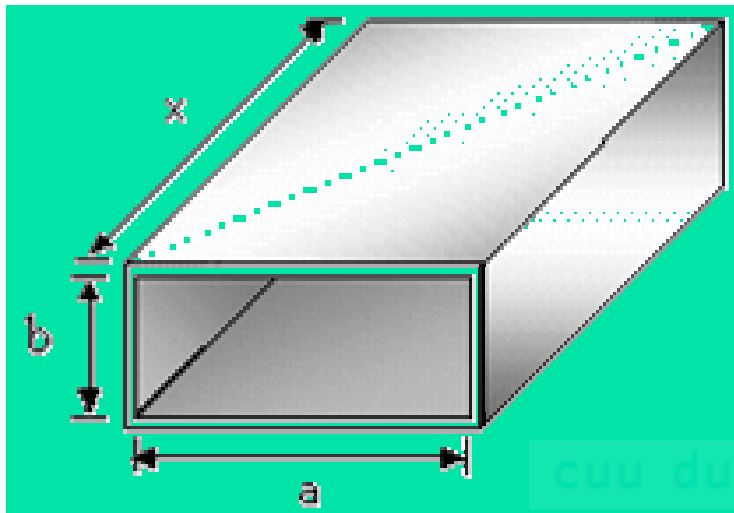


c) Circular waveguide

❖ Microstrip line :



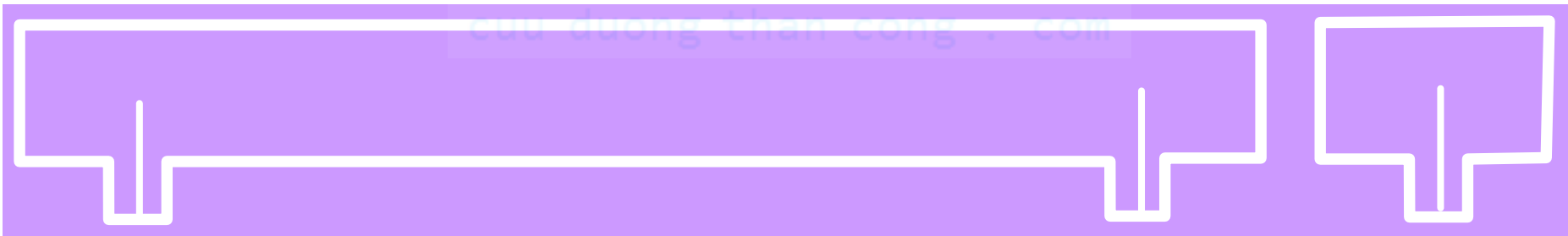
❖ Rectangular Waveguide:



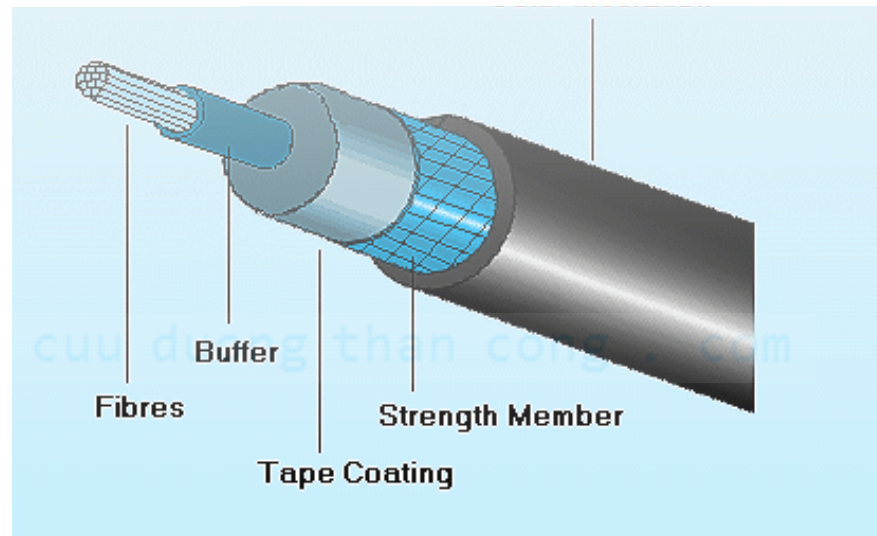
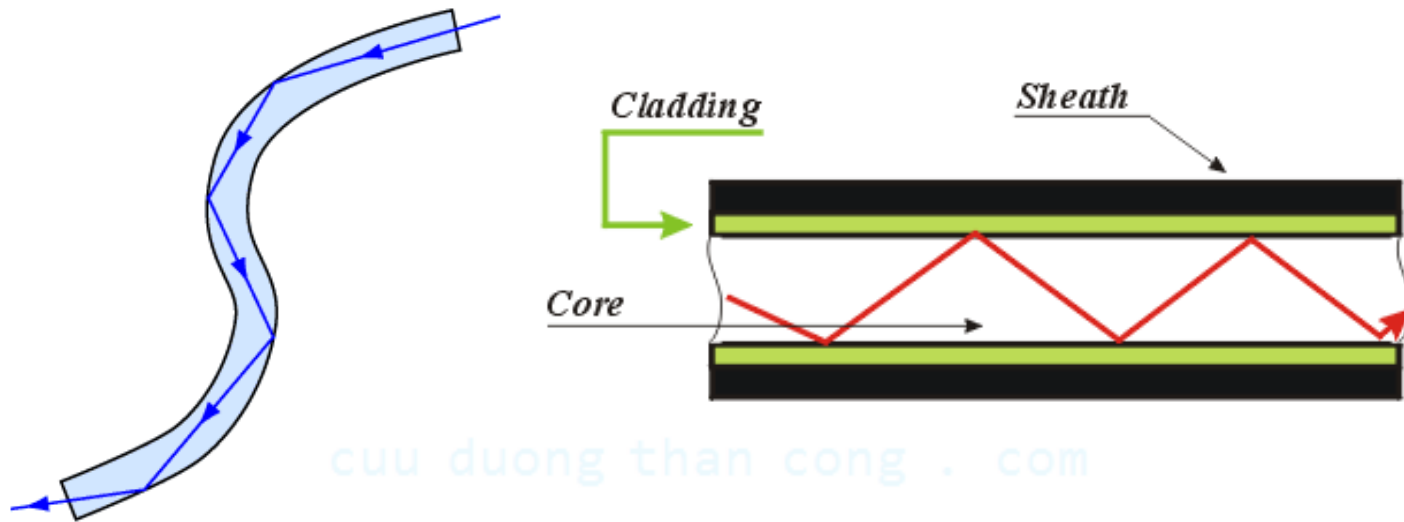
**rectangular
waveguides**



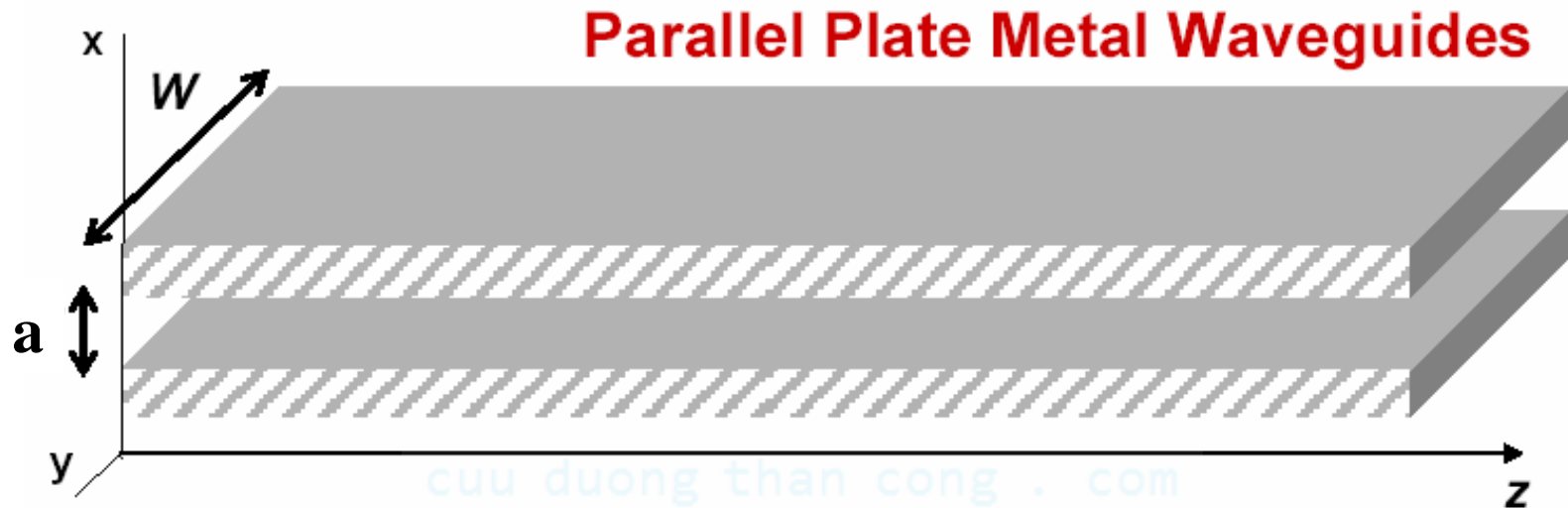
**cross
section**



❖ Optical Fibre:

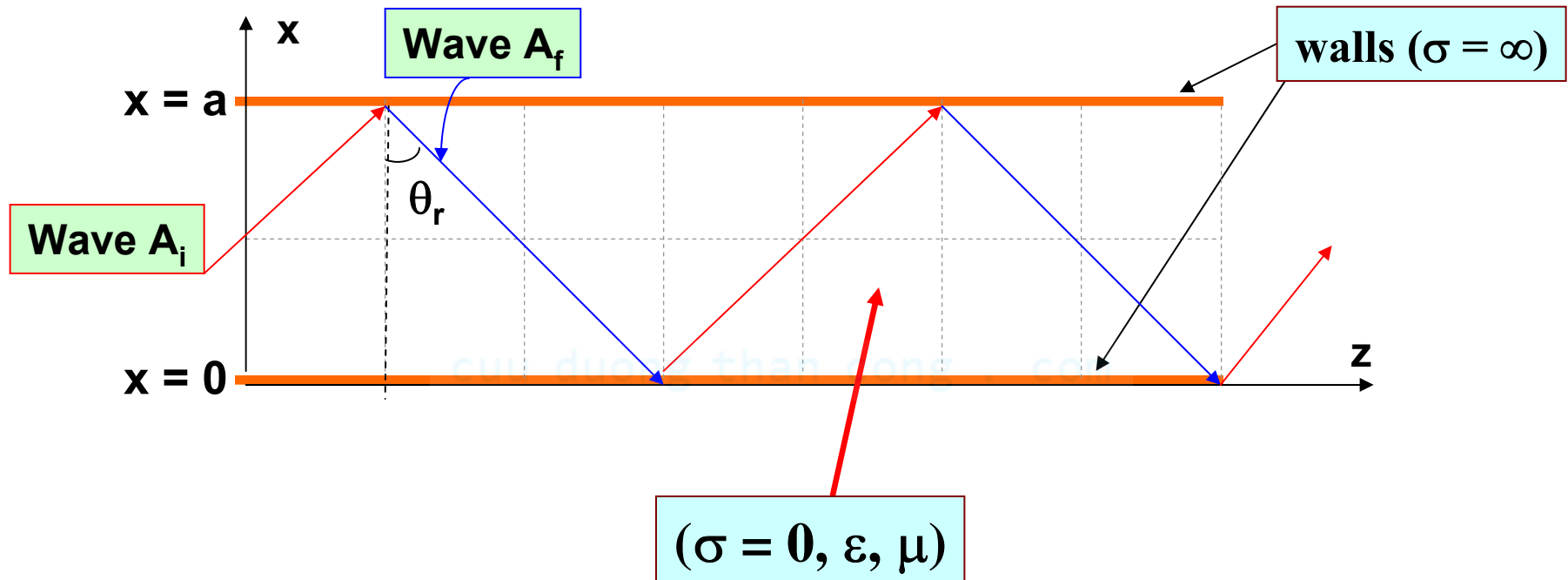


6.2: Parallel-plate waveguide:

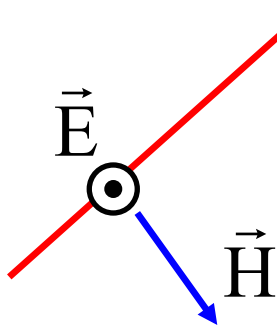
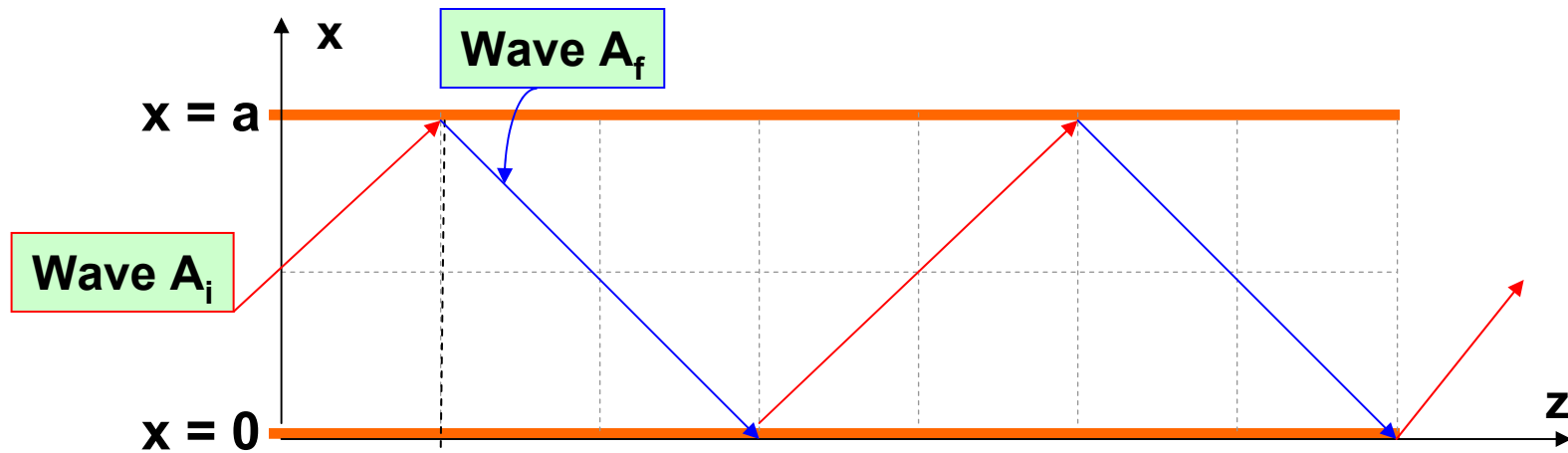


- Consider a parallel plate waveguide (shown above)
- We have studied such structures in the context of **transmission lines**
- We know that they can guide **TEM** waves (**T**ransverse **E**lectric and **M**agnetic) in which both the electric and magnetic fields point in direction perpendicular to the propagation direction
- But these structures can guide more than just the **TEM** waves that we have considered so far

a) The Principle:



b) The Guide Modes:

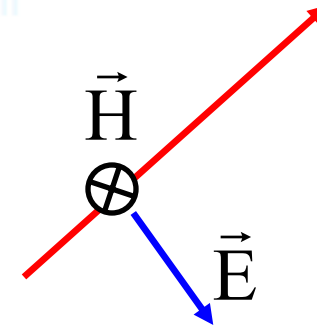


$$\vec{E} = E_y \vec{a}_y$$

$$\vec{H} = H_x \vec{a}_x + H_z \vec{a}_z$$



TE wave



$$\vec{E} = E_x \vec{a}_x + E_z \vec{a}_z$$

$$\vec{H} = H_y \vec{a}_y$$



TM wave

c) The TM Guidewave Modes :

❖ Assume sinusoidal field. In phasor form :

$$\dot{\vec{E}}_x = E_x(x).e^{-\gamma z} \quad ; \quad \dot{\vec{E}}_z = E_z(x).e^{-\gamma z}$$

$$\dot{H}_y = H_y(x).e^{-\gamma z}$$

❖ Maxwell's equations :

$$\text{rot } \vec{H} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \partial/\partial x & 0 & \partial/\partial z \\ 0 & \dot{H}_y & 0 \end{vmatrix} = j\omega\epsilon \vec{E} \quad \Rightarrow \quad \begin{cases} \gamma \dot{H}_y = j\omega\epsilon \dot{E}_x & (1) \\ \frac{\partial \dot{H}_y}{\partial x} = j\omega\epsilon \dot{E}_z & (2) \end{cases}$$

$$\text{rot } \vec{E} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \partial/\partial x & 0 & \partial/\partial z \\ \dot{E}_x & 0 & \dot{E}_z \end{vmatrix} = -j\omega\mu \vec{H} \quad \Rightarrow \quad -\gamma \dot{E}_x - \frac{\partial \dot{E}_z}{\partial x} = -j\omega\mu \dot{H}_y \quad (3)$$

❖ The component $\dot{\mathbf{E}}_z$:

$$\dot{\mathbf{E}}_x = \frac{\gamma}{j\omega\epsilon} \dot{\mathbf{H}}_y \quad \longrightarrow \quad \frac{\partial \dot{\mathbf{E}}_z}{\partial x} + \left(\frac{\gamma^2}{j\omega\epsilon} - j\omega\mu \right) \dot{\mathbf{H}}_y = 0$$

$$\longrightarrow \frac{\partial^2 \dot{\mathbf{E}}_z}{\partial x^2} + (\gamma^2 + \omega^2 \mu \epsilon) \dot{\mathbf{E}}_z = 0 \quad (\gamma = j\beta_z)$$

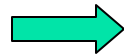
$$\longrightarrow \frac{\partial^2 \dot{\mathbf{E}}_z}{\partial x^2} + K_c^2 \dot{\mathbf{E}}_z = 0$$

$$(K_c^2 = \omega^2 \mu \epsilon + \gamma^2 = \omega^2 \mu \epsilon - \beta_z^2)$$



❖ The other components :

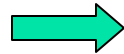
$$-\frac{\partial \dot{E}_z}{\partial x} = \left(\gamma + \frac{\omega^2 \mu \epsilon}{\gamma} \right) \dot{E}_x$$



$$\dot{E}_x = -\frac{\gamma}{K_C^2} \frac{\partial \dot{E}_z}{\partial x}$$

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$$\dot{E}_x = \frac{\gamma}{j\omega\epsilon} \dot{H}_y$$

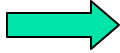


$$\dot{H}_y = \frac{\dot{E}_x}{\eta_{TM}}$$

$$(\eta_{TM} = \frac{\beta_z}{\omega\epsilon})$$

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❖ The TM_m wave :


$$\dot{E}_Z = C \sin\left(\frac{m\pi}{a}x\right) \cdot e^{-\gamma Z}$$

$m = 0, 1, 2, 3 \dots$
(mode index)

(Note: $m = 0$: TEM wave)

$$\dot{E}_x = -\frac{\gamma}{K_c^2} \frac{\partial \dot{E}_z}{\partial x}$$

$$\dot{H}_y = \frac{\dot{E}_x}{\eta_{TM}}$$

$$K_c^2 = \left(\frac{m\pi}{a}\right)^2$$

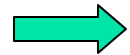
$$\gamma = j\beta_z$$

$$\beta_z = \sqrt{\left(\frac{\omega}{v}\right)^2 - \left(\frac{m\pi}{a}\right)^2}$$

$$(\eta_{TM} = \frac{\beta_z}{\omega\epsilon})$$

d) The TE_m wave :

$$\frac{\partial^2 \dot{H}_z}{\partial x^2} + K_c^2 \dot{H}_z = 0$$



$$\dot{H}_z = C \cos\left(\frac{m\pi}{a} x\right) \cdot e^{-\gamma z}$$

m = 1, 2, 3 ...

(mode index)

(Note: m = 0 : H_x & E_y = 0)

$$\dot{H}_x = -\frac{\gamma}{K_c^2} \frac{\partial \dot{H}_z}{\partial x}$$

$$\dot{E}_y = -\eta_{TE} \cdot \dot{H}_x$$

$$\left(\eta_{TE} = \frac{\omega \mu}{\beta_z} \right)$$

e) Summary :

i. Cutoff-frequency:

$$f > \frac{v}{2} \cdot \frac{m}{a} = f_c$$

$$\lambda = \frac{v}{f} < \frac{2a}{m} = \lambda_c$$

ii. Phase constant:

$$\beta_z = \sqrt{\left(\frac{\omega}{v}\right)^2 - \left(\frac{m\pi}{a}\right)^2} = \beta \sqrt{1 - (f_c / f)^2}$$

❖ If $f < f_c$: $\gamma^2 = (m\pi / a)^2 - \omega^2 \mu \epsilon > 0$

➡ $\alpha_z = \sqrt{\left(\frac{m\pi}{a}\right)^2 - \left(\frac{\omega}{v}\right)^2}$ ➡ **Attenuation !!!**

iii. Velocity :

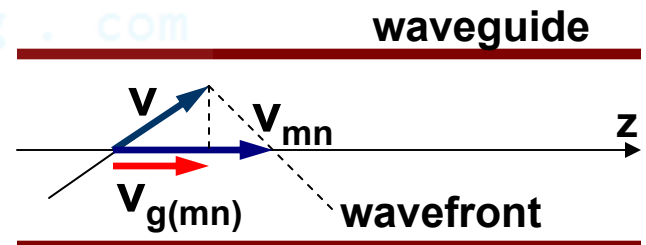
$$V_m = \frac{\omega}{\beta_z} = \frac{v}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}} \geq v$$

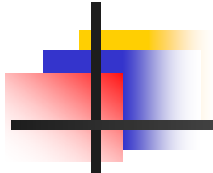
❖ Nhận xét:

- a) Vận tốc pha trong ods luôn lớn hơn vận tốc truyền sóng trong không gian tự do .
- b) Khi truyền tín hiệu , TĐT biến thiên sẽ bị điều chế để mang theo thông tin. Lúc này , vận tốc của nhóm sóng (vận tốc của các mặt phẳng vuông góc trục z) là cần thiết :

Và người ta chứng minh được :

$$V_{g(mn)} \cdot V_{mn} = v^2$$





iv. Wavelength:

$$\lambda_m = \frac{2\pi}{\beta_z} = \frac{\lambda}{\sqrt{1 - (f_c / f)^2}}$$

v. Wave impedance:

$$\eta_{TE} = \frac{\eta}{\sqrt{1 - (f_c / f)^2}}$$

$$\eta_{TM} = \eta \sqrt{1 - (f_c / f)^2}$$